

A Simple Wideband Passive Scatterer Reducing a Corner Diffraction Loss

Roshanak Zabihi* Christopher G. Hynes Rodney G. Vaughan

Sierra Wireless Communications Lab
Simon Fraser University, Canada

URSI GASS 2020

Rome, Italy
August 2020

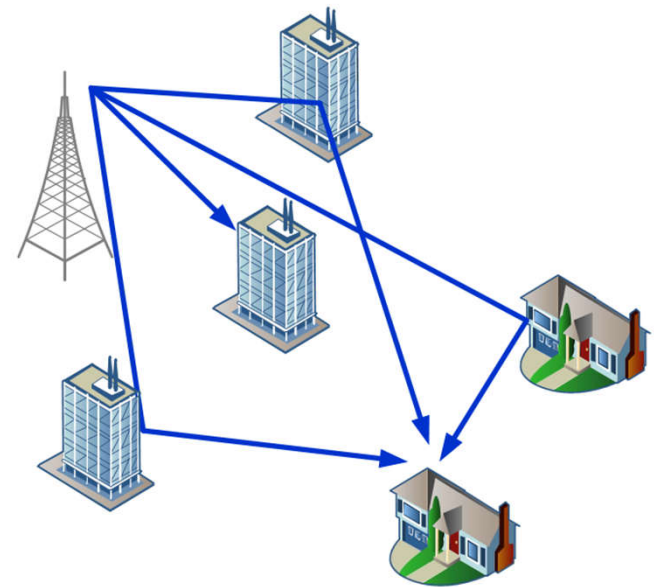
Outline

- ❑ Introduction
- ❑ Description of the Model used for Demonstration
- ❑ Simulation Approach
- ❑ Diffraction by a Corner – Uniform Theory of Diffraction
- ❑ Results
- ❑ Conclusions

Introduction

- ❑ Increasing **carrier frequencies** for communication capacity brings **greater shadowing** compromising the radio coverage.
- ❑ mmWave signals are severely attenuated in the simplest of non-line-of-sight (**NLOS**) scenarios - **around the corner of a building**.

Diffraction is a dominant propagation mechanism.



Introduction

- ❑ A traditional method to increase signal coverage in diffraction situations is the use of **passive repeaters**.
- ❑ **Passive repeaters** are a very **low cost** way to “**fill in**” a **shadow region** compared to **active repeaters*** having an amplifier stage or fully regenerate the signal.
- ❑ Examples of passive repeaters include Yang’s double flat reflectors from 1957*, Norton’s large passive reflectors from 1962**, and more recently, using Yagi-Uda antennas*** and four-element patch antenna arrays****.
- ❑ There is ongoing interest within the communications community in the concept of **adaptive, active scattering surfaces** to improve NLOS link gains**.

* R. Yang. Passive repeater using double flat reflectors. In *1958 IRE International Convention Record*, volume 5, pages 36–41, March 1957.

** M. Norton. Microwave system engineering using large passive reflectors. *IRE Transactions on Communications Systems*, 10(3):304–311, Sep. 1962.

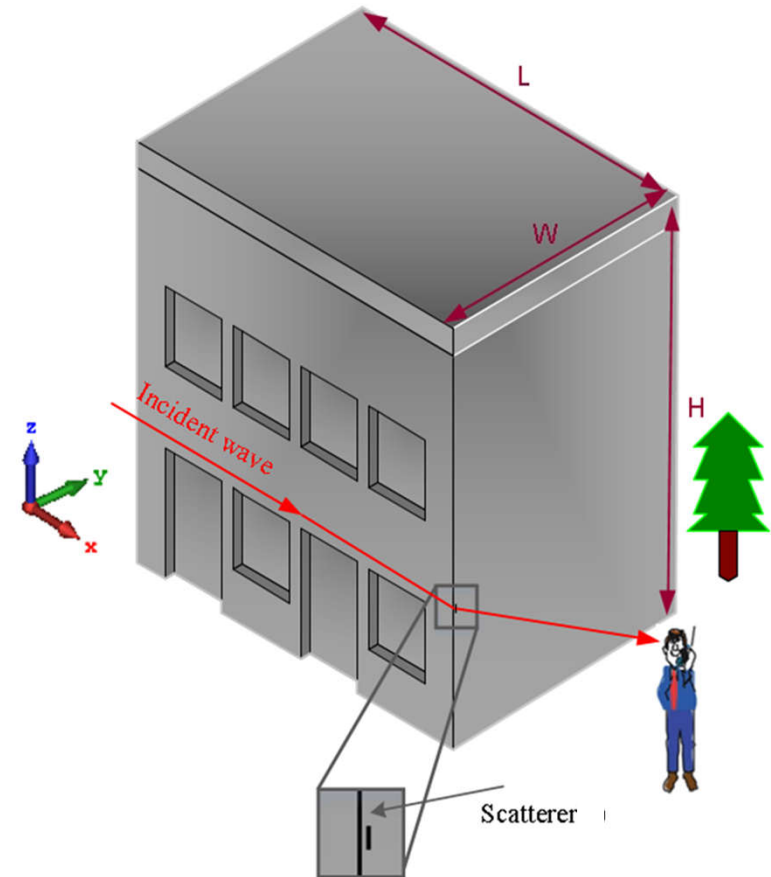
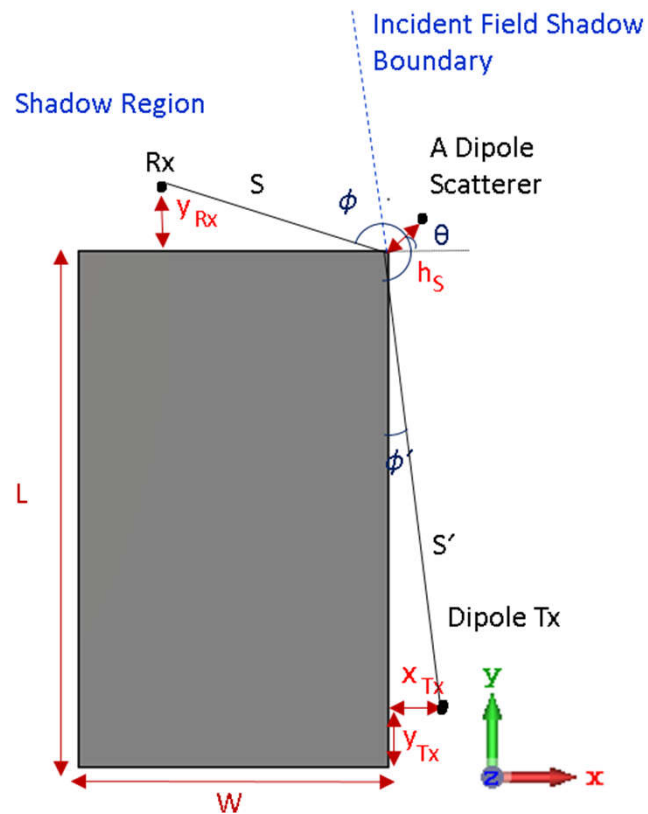
*** J. I. Chaves, A. G. Junior, and J. R. Descardec. Design of passive retransmitting system for cellular communication. volume 2, pages 76–83, Dec. 2001.

**** T. Tang, T. Hong, C. Liu, W. Zhao, and M. Kadoch. Design of 5G dual-antenna passive repeater based on machine learning. In *2019 15th International Wireless Communications Mobile Computing Conference (IWCMC)*, pages 1907–1912, June 2019.

* J. Oh, M. Thiel, and K. Sarabandi. Wave-propagation management in indoor environments using micro-radio-repeater systems. *IEEE Antennas and Propagation Magazine*, 56(2):76–88, April 2014.

** E. Basar, M. Di Renzo, J. De Rosny, M. Debbah, M. Alouini, and R. Zhang. Wireless communications through reconfigurable intelligent surfaces. *IEEE Access*, 7:116753–116773, 2019.

Description of the Model used for Demonstration

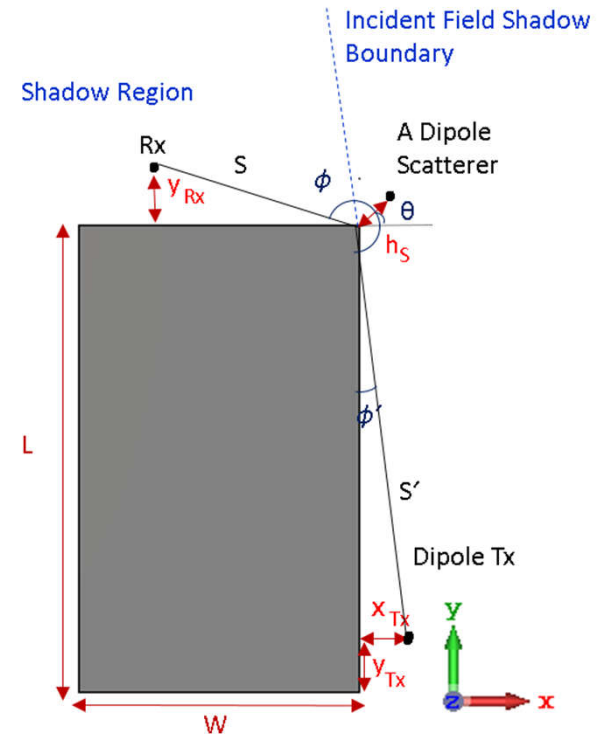


The geometry of the model (top view) used for simulation and measurement.

Propagation path around a corner.

Description of the Model used for Demonstration

- ❑ A simple **dipole scatterer** placed on the corner.
- ❑ The polarization is **vertical** (parallel case).
- ❑ The receiver is at $Y_{RX} = 2\lambda$ from the wall moved along the x-axis.
- ❑ The spacing between the dipole and the corner is $h_s = 0.8\lambda$ with $\theta = 45^\circ$.
- ❑ The receiver, scattering dipole, and the transmitter are **aligned** in the plane of incidence.
- ❑ The transmit dipole is located $Y_{TX} = X_{TX} = \lambda$.



Dimension of the model	
W	6λ
L	10λ
H	14λ

Simulation Approach*

- ❑ The CST Microwave Studio **time domain** solver is used for simulation.
- ❑ The simulation boundaries are defined as **open boundaries** with **perfectly matched layer** (PML).
- ❑ We use a **dipole source** because of the **energy leakage** problem from using an open boundary with plane wave illumination.
- ❑ There is also leakage through any dielectric material with dipole illumination, so we use **perfectly electrical conductor** (PEC) for the building, which is a limitation in our results.

* C. G. Hynes, R. Zabihi, and R. G. Vaughan. 3D simulation of infinite baffle diffraction. *Accepted in 14th European Conference on Antennas and Propagation, EuCAP*, March 2020.

Diffraction by a Corner - UTD

The critical and complex 2D uniform theory of diffraction (UTD) equations*

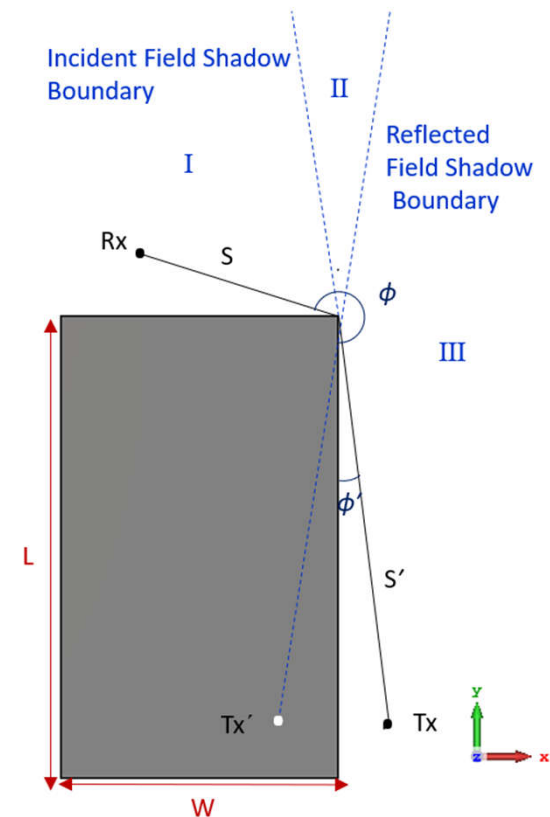
$$E_{UTD} = E_i D(\phi', \phi) A(s) e^{-jks}$$

$$D(\phi', \phi) = -\frac{e^{-j\pi/4}}{2n\sqrt{2\pi k}} \times \left[\cot\left(\frac{\pi + (\phi - \phi')}{2n}\right) F(k\mathcal{L}a^+(\phi - \phi')) + \cot\left(\frac{\pi - (\phi - \phi')}{2n}\right) F(k\mathcal{L}a^-(\phi - \phi')) + R_1 \cot\left(\frac{\pi - (\phi + \phi')}{2n}\right) F(k\mathcal{L}a^-(\phi + \phi')) + R_2 \cot\left(\frac{\pi + (\phi + \phi')}{2n}\right) F(k\mathcal{L}a^+(\phi + \phi')) \right]$$

$$a^\pm = 2 \cos^2\left(\frac{2\pi n N^\pm - \beta}{2}\right),$$

$$\beta = \phi \pm \phi',$$

$$N^\pm = \frac{\pm\pi + \beta}{2\pi n}.$$

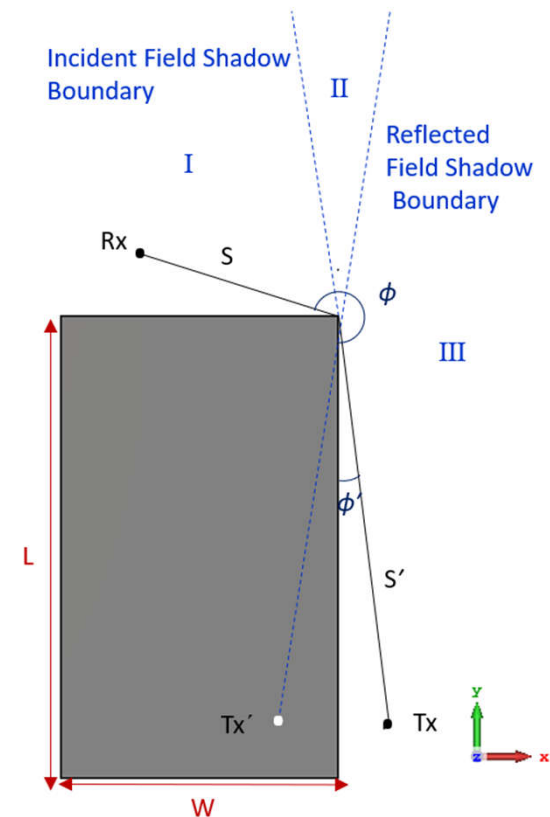


*R. G. Kouyoumjian and P. H. Pathak, "A uniform geometrical theory of diffraction for an edge in a perfectly conducting surface," *Proceedings of the IEEE*, vol. 62, no. 11, pp. 1448–1461, Nov 1974.

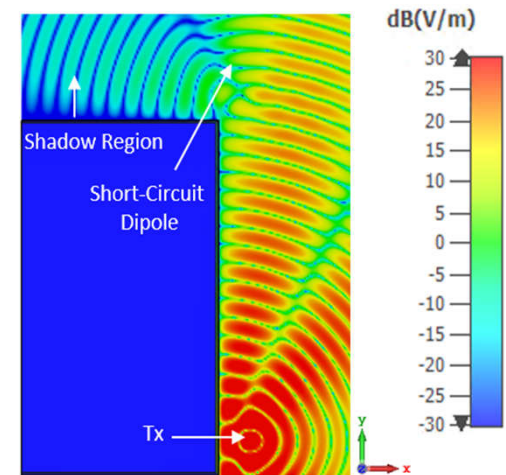
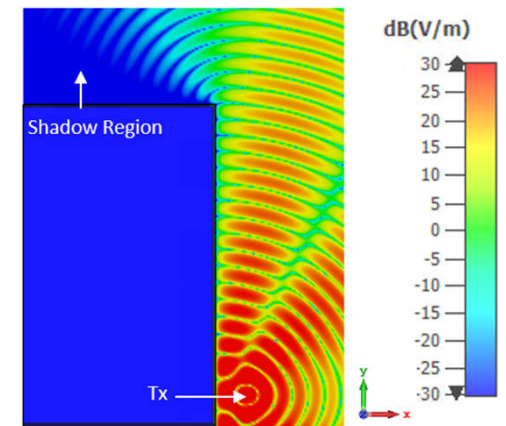
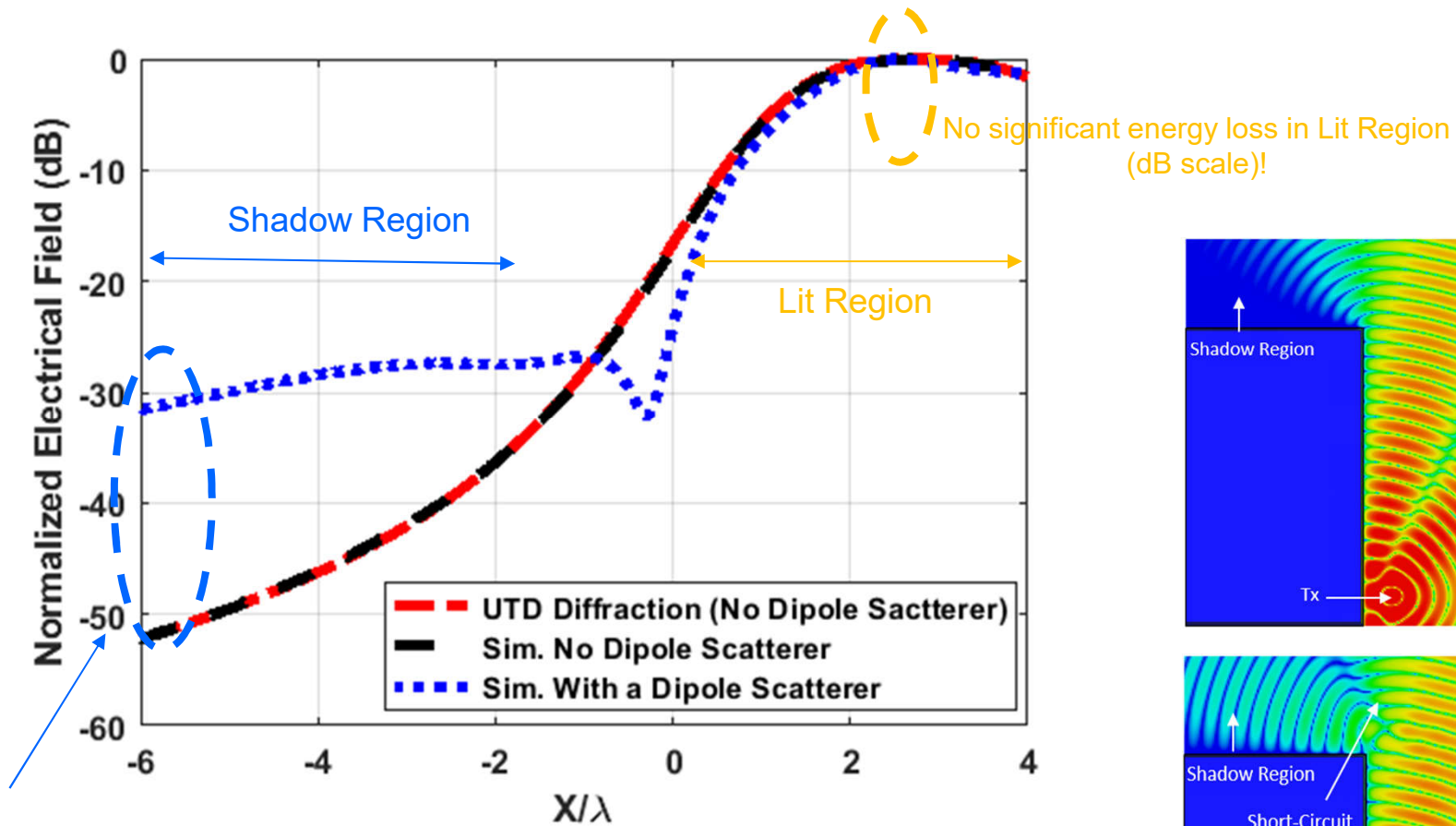
Diffraction by a Corner - UTD

- The reflection coefficients of each facet are R1 (lit side) and R2 (shadow side).

Incident field	$E_i = E_0 \frac{e^{-jks'}}{s'}$
Spreading factor	$A(s) = \sqrt{\frac{s'}{s(s' + s)}}$
Distance factor	$\mathcal{L} = \frac{ss'}{s + s'}$
Transition function	$F(x) = 2j\sqrt{x}e^{jx} \int_{\sqrt{x}}^{\infty} e^{-ju^2} du$



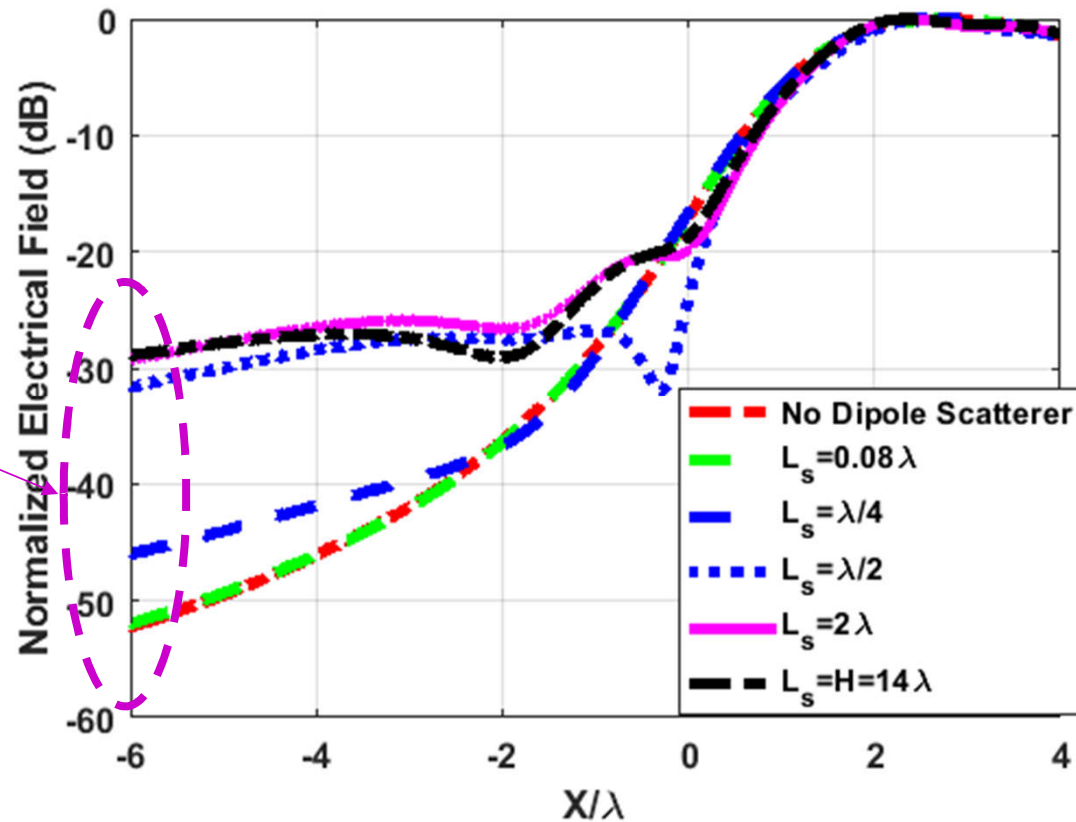
Parallel Dipole Scatterer over the Corner



The enhancement of ~20dB in the Shadow Region.

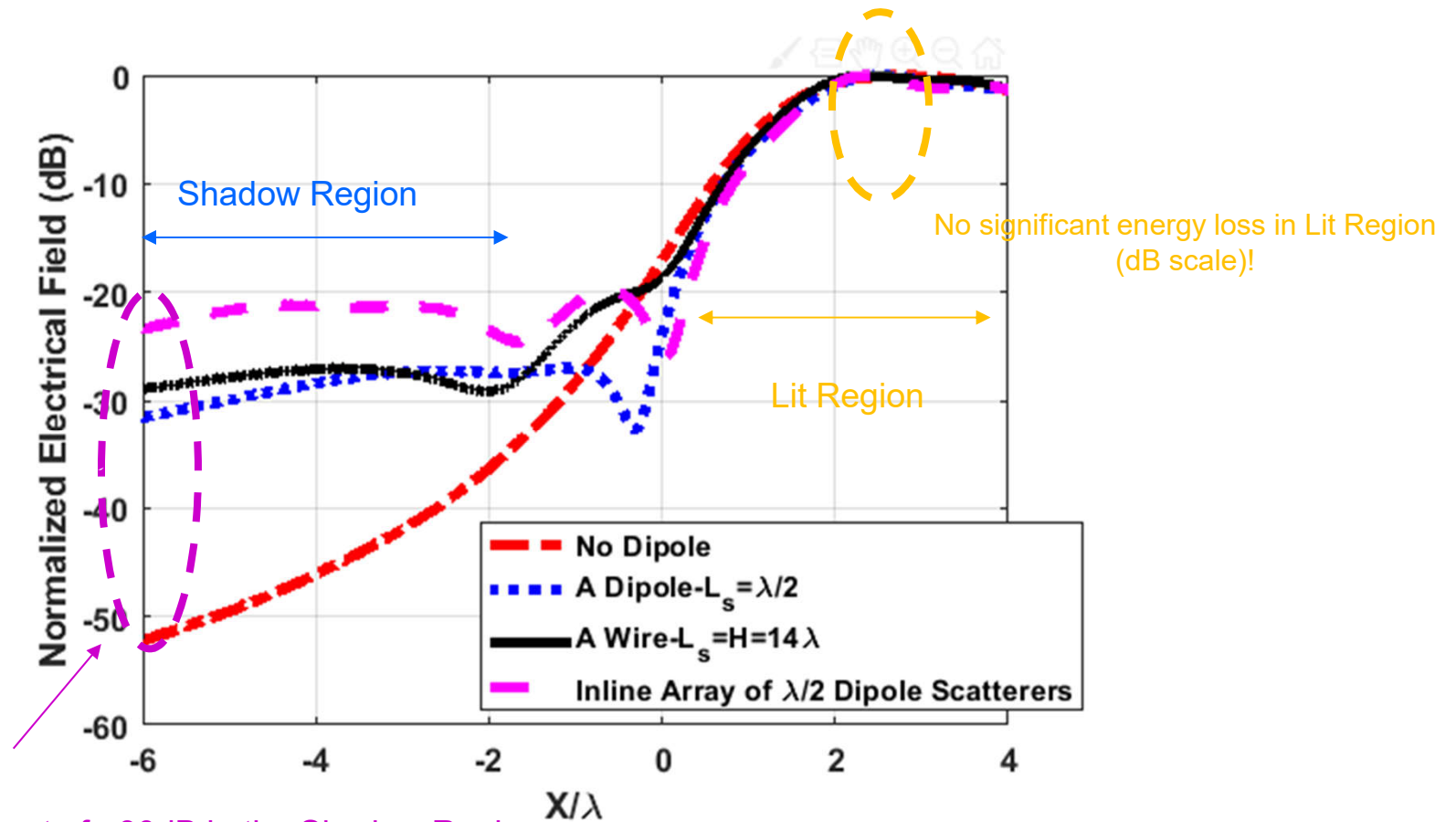
Is It Wideband?

By changing the electrical size of the dipole (L_s), a large gain is obtained from 1 to 28!



A large gain is available in the immediate shadow region over a wide range of frequencies.

Impact of Inline Array of Dipole Scatterers

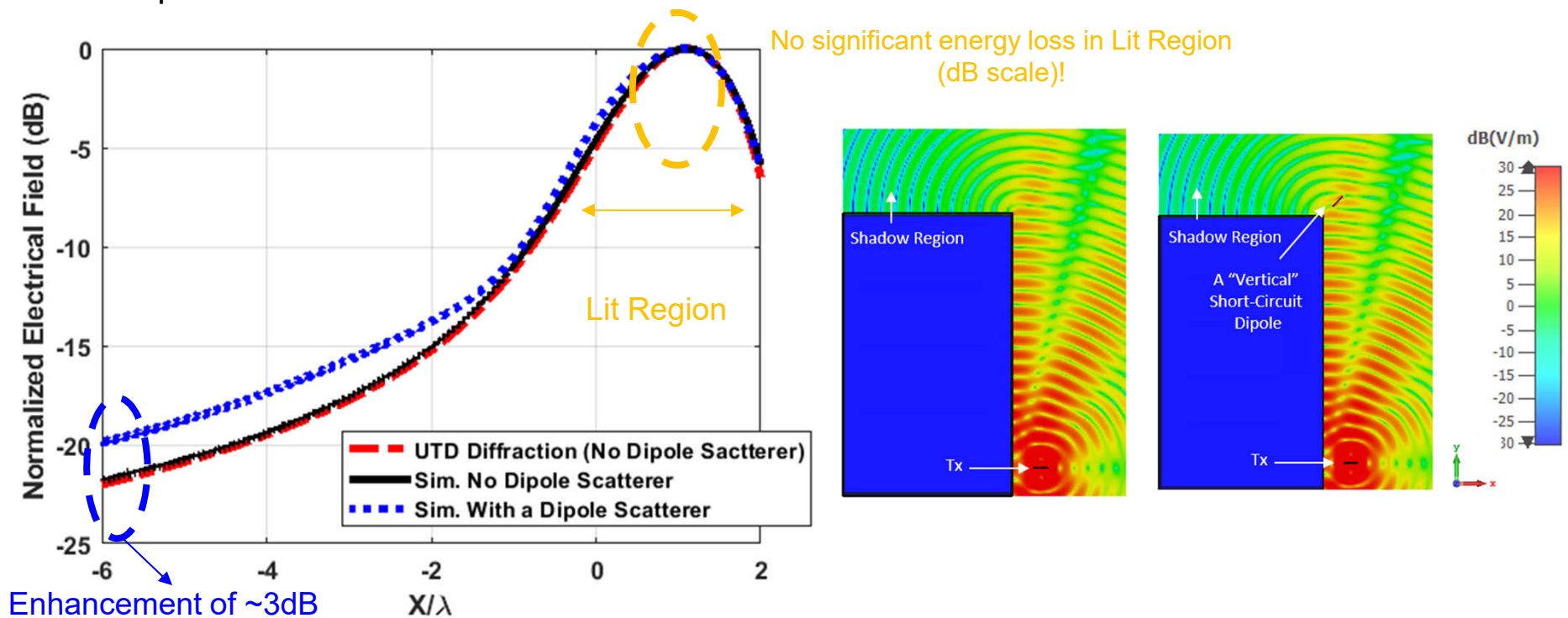


The enhancement of ~30dB in the Shadow Region.

A larger scattering aperture is of obvious interest, and with an eye to array gain and beamforming possibilities using array techniques.

Perpendicular Dipole over the Corner

- The improvement is not very significant for this case (although for aspects, such as coding and modulation, a couple of dB is a large improvement):
 - The corner diffraction attenuation for horizontal (perpendicular) polarization is less than that of the vertical (parallel) polarization.
 - The diffraction attenuation problem is not as important as for the vertical polarization case.



Summary

- ❑ We present a **simple corner modification** comprising a **fixed scattering dipole** to improve the coverage in the **shadow region** of the **corner diffraction**.
- ❑ The UTD wedge diffraction formulation offers a benchmark for the shadowing, but this is complicated to simulate using CST Microwave Studio. We describe our simulation approach and demonstrate a match between the **simulation** and **diffraction** results, despite using a dipole excitation instead of a plane wave.
- ❑ For the **vertical polarization**, a signal level increase of well **over ten dB** in the shadow region is demonstrated using a single half-wavelength dipole scatterer. While **arrays of dipole scatterers** can further improve the energy enhancement
- ❑ The scattering dipole is shown to have a strong effect in the deep shadow region, and little effect in the lit region, for vertical (parallel) polarization. Smaller signal enhancements are shown for horizontal (perpendicular) polarization.
- ❑ As noted, a **limitation** of our analysis is that the building is **conducting**. Most real-world buildings are not conducting, but the close proximity scatterer means that the reflections off the building surfaces are at grazing angles. This in turn means that the **surface detail** (conductivity or roughness) **is not critical** for the dipole scattering mechanism to work well.

Further question can be sent to

Roshanak Zabihi - rzabihi@sfu.ca

Christopher G. Hynes - ch@sfu.ca

Rodney G. Vaughan - rodney_vaughan@sfu.ca