

Centre d'Investigació en Metamaterials per a la Innovació en Tecnologies Electrònica i de Comunicacions

### Chipless-RFID Sensors for Motion Control Applications

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URSI GASS 2020, September 2020, Roma (Italy)

UAB Research Park



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- 1. Motivation & Objectives
- 2. Working Principle
- 3. Reader and Encoder
- 4. Fabrication and Measurement
- 5. Conclusion



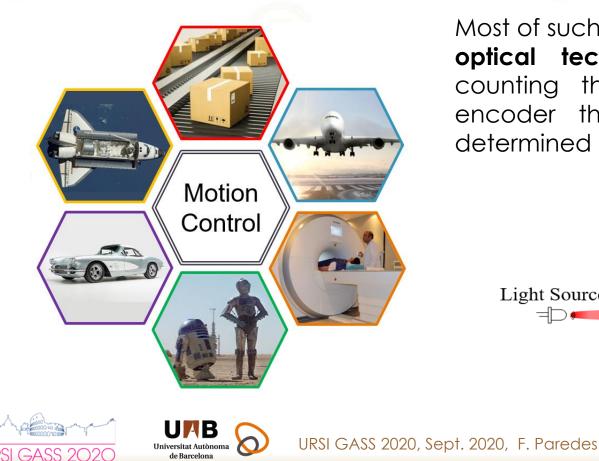




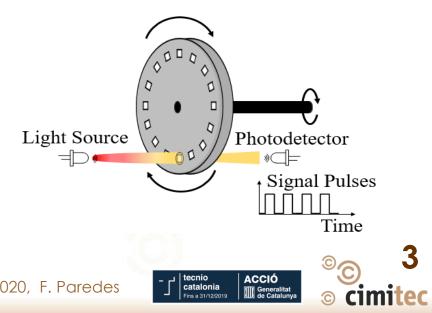


### 1. Motivation & Objectives

In **Motion Control applications** it is often required to use a huge amount of sensors in order to determine velocity, acceleration, displacement, among others magnitudes.



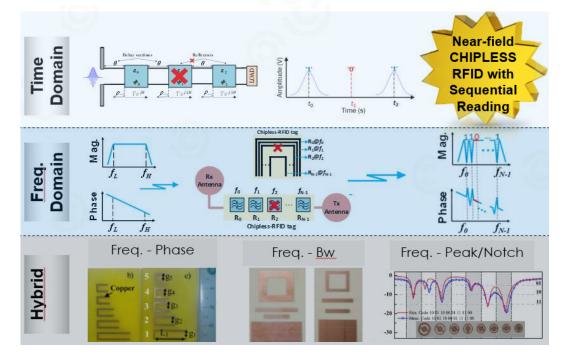
Most of such sensors are based on optical technology, where by counting the apertures in the encoder the position can be determined (incremental position)



# 1. Motivation & Objectives

The main objective of this work is to implement a **microwave sensor**, to measure displacement & velocity/acceleration, based on the functionality of optical encoders.

The system is inspired by the Near-Field Chipless-RFID approach.



The Microwave Sensor:

- Designed on a PCB (planar technology)

- Provide the absolute position (Bruijn)







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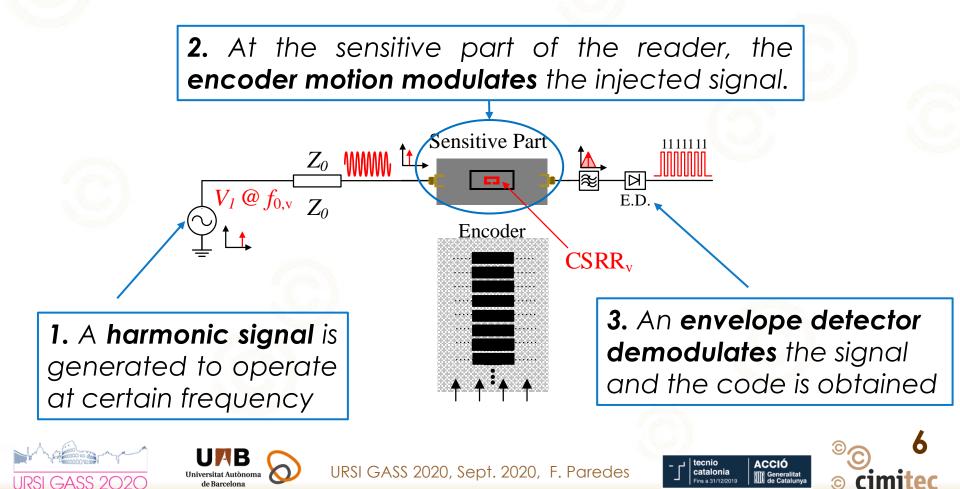






# 2. Working Principle

The proposed system has the same **working principle** that **Near-Field Chipless-RFID:** 

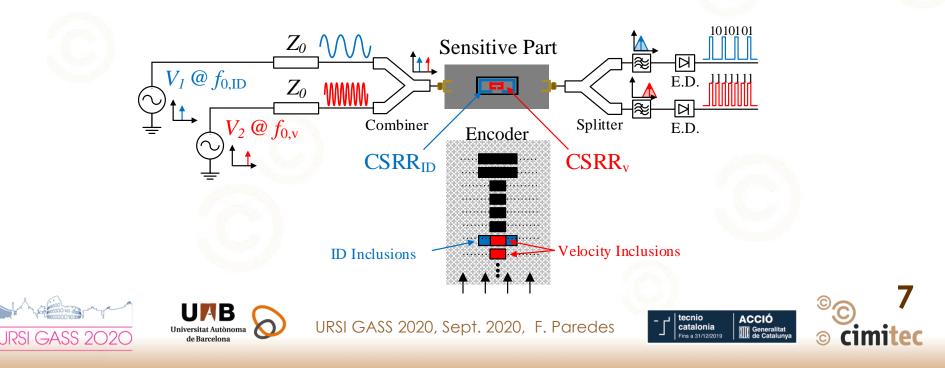


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### 2. Working Principle

For Synchronous reading it is required to have at least a pair of harmonic signals ( $V_1 \& V_2$  pure tones)  $\rightarrow$  Therefore, two sensing elements in the sensitive part of the reader are needed.

On the **encoder**, only **one chain of metallic patches** is employed in this approach and the binary states can be distinguished according the **patch size (width)**.



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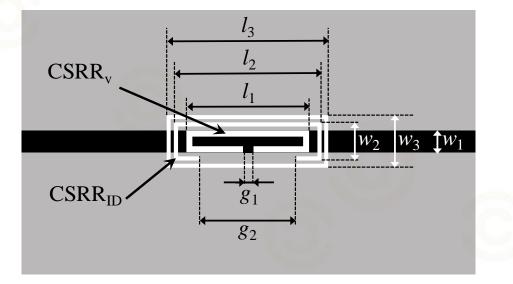


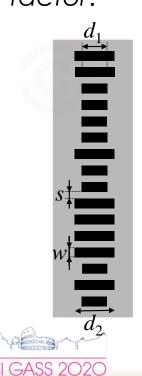






The **Sensitive Part** of the reader consists of Microstrip Transmission Line loaded with **two CSRRs**, and an outer ring for tailoring the Q factor.







respectively.

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The **Encoder** is made of **metallic patches**.

purposes, as well as to obtain the **velocity**.

The **presence** of the patch is used for **synchronous** 

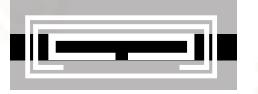
The **binary states** are associated to the **patch size**. The

wider and narrow patches correspond to '1' and '0',

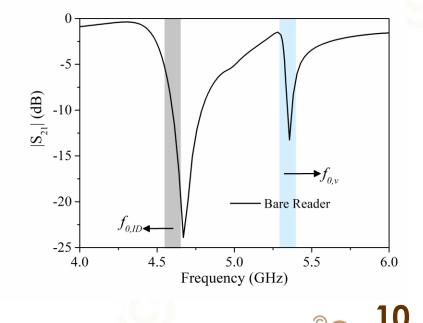




#### Sensitive Part



The Sensitive Part of the reader has **two operation frequencies** 



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#### Sensitive Part

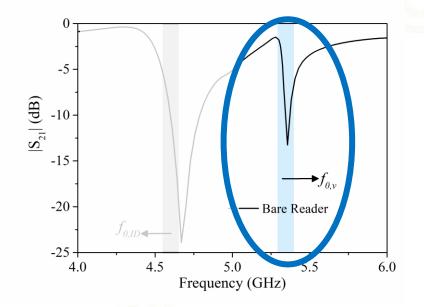




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The Inner CSRR provides the higher resonant frequency at:

@  $f_{0,v} = 5.31 \text{ GHz}$ 





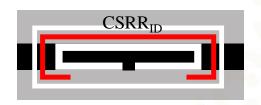




#### Sensitive Part

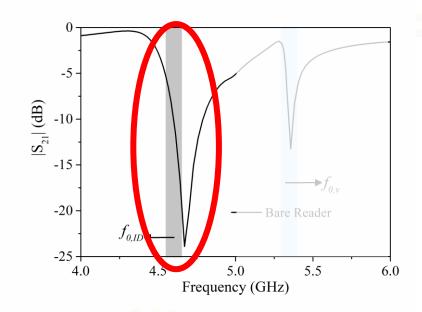






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The Outer CSRR provides the lower resonant frequency at:  $@ f_{\theta, ID} = 4.63 \text{ GHz}$ 





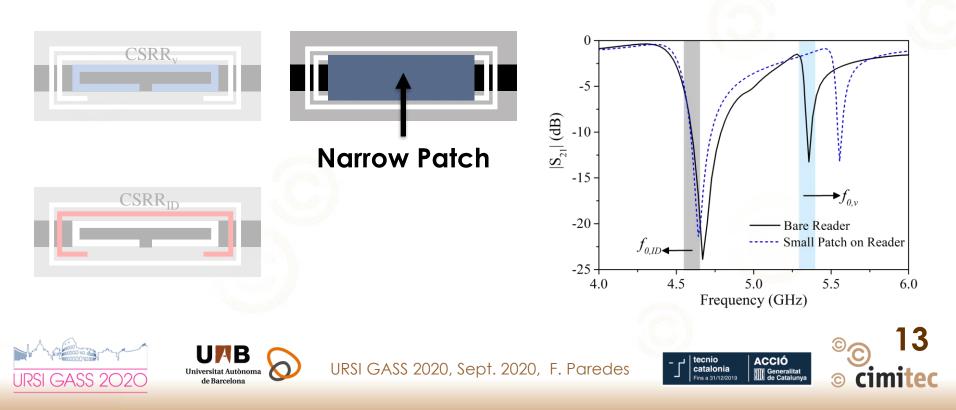




#### **Sensitive Part**



When a patch relies on top of the Sensitive Part The **Narrow** Patch only has effect in the inner CSRR, resulting in a shift of the **higher** frequency

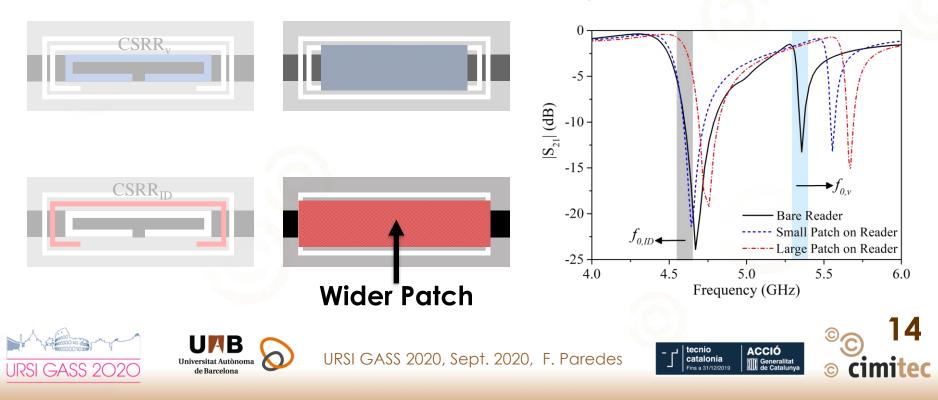




#### **Sensitive Part**



When a patch relies on top of the Sensitive Part The Wider Patch has effect on both the inner and the outer CSRRs, resulting in a shift of the **lower** and the **higher** frequencies



#### $@f_{\theta,ID}$ $@f_{\theta,v}$ Bare Reader '0' '0' SUMMARY of the **Binary States** '1' Narrow Patch '0' Wider Patch '1' '1' Wider Patch **Bare Reader** Narrow Patch -5 -5 -5 $\frac{(R)^{-10}}{(R)^{-10}}$ $\frac{|S_{21}|}{|S_{21}|}$ (dB) $|S_{2l}|$ (dB) -10 -15 -15 -15 → f<sub>0,v</sub> -20 -20 Bare Reader -20 Bare Reader Bare Reader ---- Small Patch on Reader mall Patch on Reader Join. JOID arge Patch on Reader -25 --25 -25 4.0 4.5 5.0 5.5 6.0 4.0 4.5 5.0 5.5 4.5 5.0 6.0 4.0 5.5 6.0 Frequency (GHz) Frequency (GI1z) Frequency (GI1z) 5 В tecnio ACCIÓ catalonia

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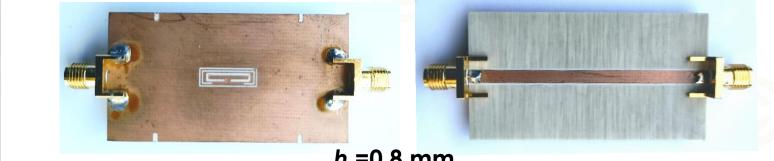






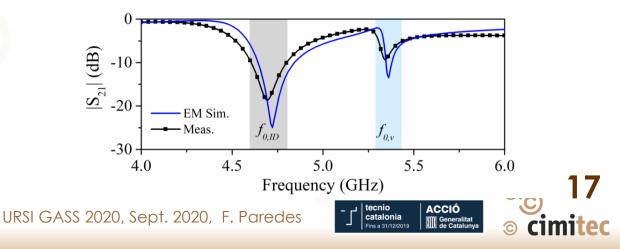


Fabricated in **Rogers RO4003C** ( $\varepsilon_r = 3.38$ , tan $\delta = 0.0022$ ) by means of a milling machine (LPKF H100)



h =0.8 mm

Electromagnetic Simulation and measurement of the sensitive part of the reader (bare reader)



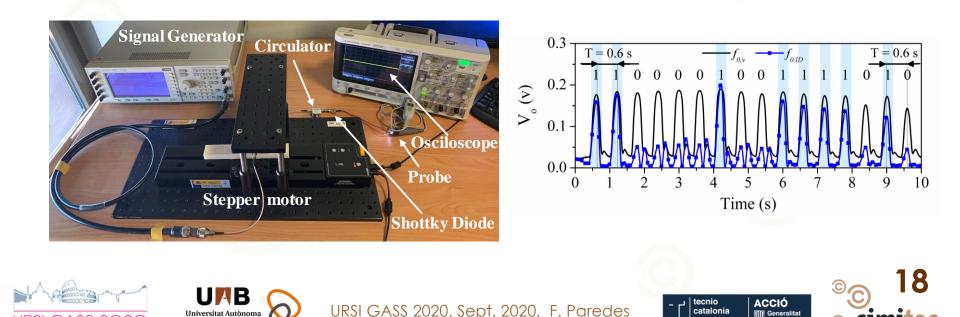
*h* =0.2 mm



The measurements were carried out **sequentially**, firstly generating the harmonic signal @  $f_{0,v}$  and later the harmonic signal @ $f_{0,D}$ .

Finally, the results were jointly **processed**.

The **black peaks** allow us to determine the encoder **velocity**, whereas the **blue peaks** are the **identification** (Bruijn sequence)



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The functionality of the system was tested by increasing the encoder **velocity** up to **14 mm/s**.

 $f_{0,ID}$ 

0

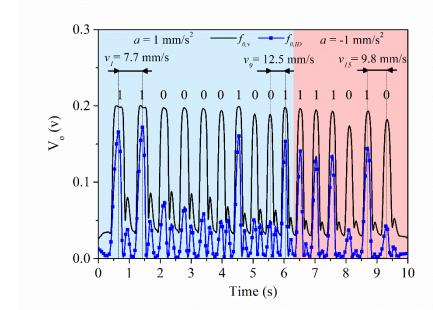
0

v = 14 mm/s

0 1 1 1 1 0

T = 0.43

The encoder was also **accelerated** at **1mm/s<sup>2</sup>**, and in the middle of the reading, the encoder was **decelerated** at **-1mm/s<sup>2</sup>**.



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0.3

0.2 -

0.1

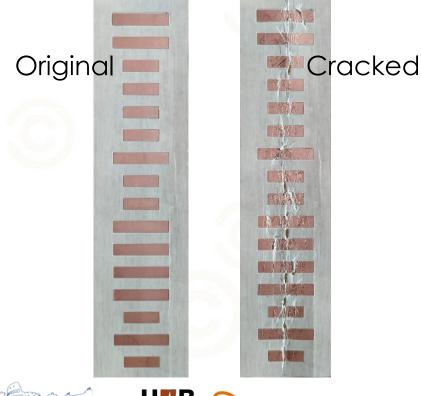
V<sub>0</sub> (v)

T = 0.43 s

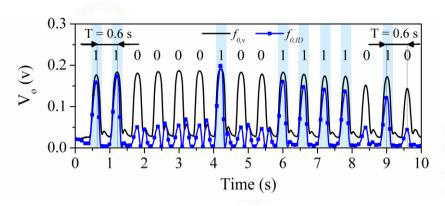


Time (s)

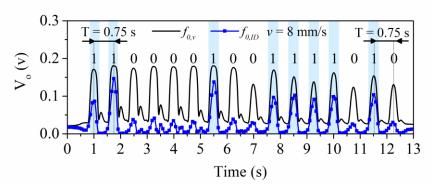
Finally, the encoder was **deliberately cracked** in order to test the robustness against **wearing or / and friction**.



#### Encoder **before** being cracked



### Encoder after being cracked



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### 5. Conclusion & Future Work

- An approach for an electromagnetic encoder useful for motion applications, and based on the Near-Field Chipless RFID approach, has been proposed.
- The Reader consists on a microstrip line loaded with a pair of complementary split ring resonators (CSRRs). The smaller CSRR is devoted to determinate the encoder velocity, whereas the longer CSRR is used to infer the ID code.
- Encoders are based on a single chain of rectangular patches, and the size of the patch determines the binary state of the ID..
- Experimental validation was carried out by reading a 16-bits encoder by testing different velocities and accelerations, as well as by cracking the encoder.









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### Thank you

### for your attention

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