Orientation Determination of a Scatterer Based on Polarimetric Radar Measurements

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Chipless RFID Technology

- Linear time-invariant systems
- Fully printable using ink-jet technology and conductive ink

Consequently, chipless tags do not have:

- battery
- emitting system
- memory

- back-scattering modulation
- media access control

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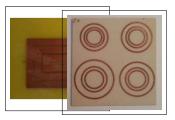
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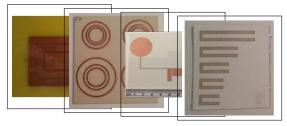
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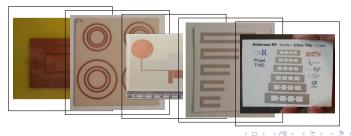
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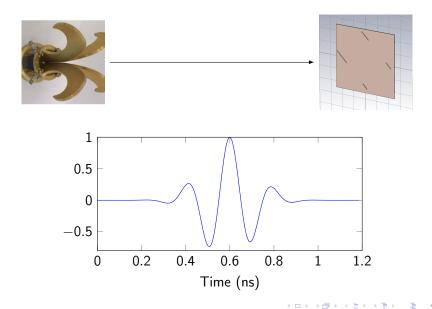
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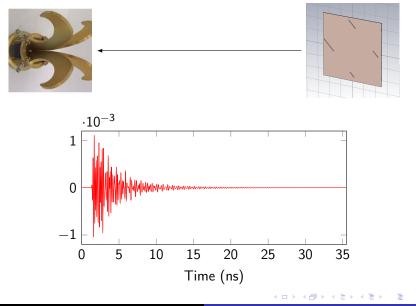
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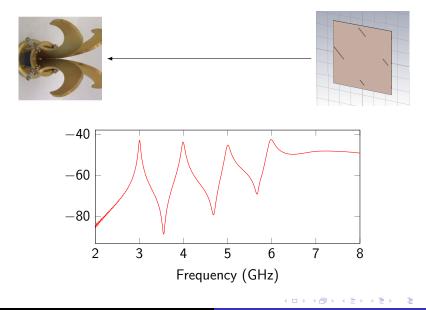


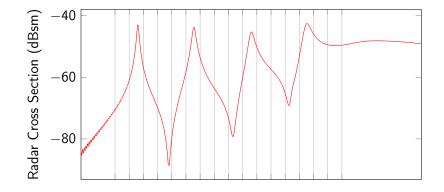


Principle



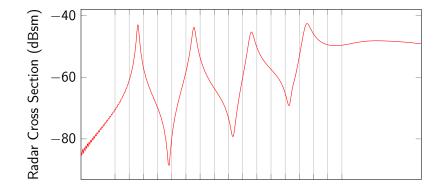
Principle





Frequency (GHz)

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Frequency (GHz)

ID: 0100010001000100

	Barcode	Chipless RFID	RFID
Discretion	-	+++	++
Writing	-	+	+++
Multiple reading	-	-	+++
Coding density	++	-	+++
Read range	+	-	+++
Sensing	-	+++	+
Cost	+++	++	-

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Objectives

Design a procedure (on the reader) side which can estimate the orientation of a chipless tag. Proposed method has the following advantages:

- valid for any scatterers
- based on an analytical model (without lookup table)
- can be used for different distances
- sensing and identification can be combined (without reducing the coding capacity)

The description of the interaction of a wave with a chipless tag is described by its polarization scattering matrix **S** which links the scattered electric field vector E^r , to the incident field vector E^i in vertical v and horizontal h polarizations:

$$\begin{bmatrix} E_{\nu}^{r} \\ E_{h}^{r} \end{bmatrix} = \begin{bmatrix} S_{\nu\nu} & S_{\nu h} \\ S_{h\nu} & S_{hh} \end{bmatrix} \cdot \begin{bmatrix} E_{\nu}^{i} \\ E_{h}^{i} \end{bmatrix}$$
(1)

- $S_{xy} \in C$ and is a function of the frequency
- $S_{vh} = S_{hv}$ since chipless tags are passive and reciprocal
- (1) is valid for any scatterer.

If we consider a rotation of the tag by an angle θ under normal incidence, we can show that the parameters of the **S** matrix at θ are linked to the initial ones by the following expression:

$$\mathbf{S}(\theta) = \mathbf{\Omega}^T \cdot \mathbf{S} \cdot \mathbf{\Omega} \tag{2}$$

where $^{\mathcal{T}}$ is the transpose operator and Ω is a rotation matrix defined by:

$$\mathbf{\Omega} = \begin{bmatrix} \cos\theta & -\sin\theta\\ \sin\theta & \cos\theta \end{bmatrix}$$
(3)

From (2), expression of $S_{\nu\nu}(\theta)$ can be written as:

$$S_{\nu\nu}(\theta) = S_{\nu\nu}\cos^2\theta + S_{\nu h}\sin 2\theta + S_{hh}\sin^2\theta \tag{4}$$

By reducing the power of \cos^2 and \sin^2 :

$$S_{\nu\nu}(\theta) = S_{\nu\nu}\frac{1+\cos 2\theta}{2} + S_{\nu h}\sin 2\theta + S_{hh}\frac{1-\cos 2\theta}{2}$$
(5)

By regrouping constant terms and $\cos 2\theta$ terms, we have:

$$S_{vv}(\theta) - \frac{S_{vv} + S_{hh}}{2} = \frac{S_{vv} - S_{hh}}{2} \cos 2\theta + S_{vh} \sin 2\theta \qquad (6)$$

Analytical Expression

Setting $\tan \theta = t$ (with $\sin 2\theta = \frac{2t}{1+t^2}$ and $\cos 2\theta = \frac{1-t^2}{1+t^2}$) and rearranging leads to:

$$S_{vv}(\theta) - \frac{S_{vv} + S_{hh}}{2} = \frac{S_{vv} - S_{hh}}{2} \frac{1 - t^2}{1 + t^2} + S_{vh} \frac{2t}{1 + t^2}$$
(7)

After developing:

$$(S_{vv}(\theta) - S_{hh})t^2 + 2S_{vh}t + (S_{vv}(\theta) - S_{vv}) = 0$$
(8)

where we can recognize a classical second order polynomial of real variable t with complex coefficients.

Analytical Expression

Solutions of this polynomial can be expressed as:

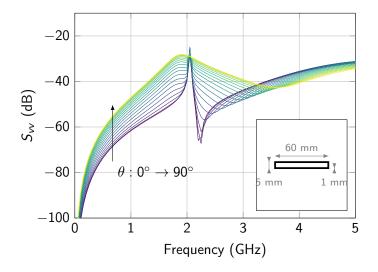
$$t_{1,2} = \frac{-2S_{\nu h} \pm \sqrt{\Delta}}{2(S_{\nu \nu}(\theta) - S_{hh})} \tag{9}$$

where $\Delta = (2S_{\nu h})^2 - 4(S_{\nu \nu}(\theta) - S_{hh})(S_{\nu \nu}(\theta) - S_{\nu \nu})$. Finally, θ can be extracted from $t_{1,2}$:

$$\hat{\theta} = \arctan(t_{1,2})$$
 (10)

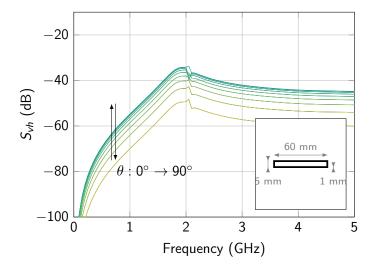
- Need 3 reference measurements: S_{vv} , S_{vh} , and S_{hh}
- 1 measurement in co-polarization $S_{\nu\nu}(\theta)$
- This determination is valid for any scatterer
- Note that $t_{1,2}$ is real (without considering noise)

Vertical Co-polarization



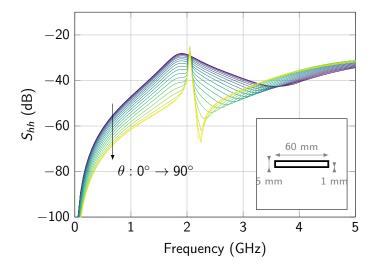
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Cross-polarization



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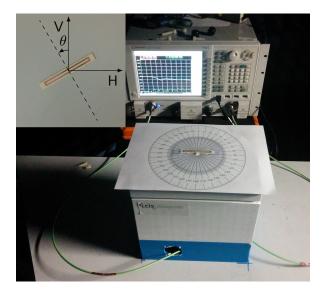
Horizontal Co-polarization



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- For horizontal loop, information is encoded into S_{vv} (sharp peak)
- S_{hh} cannot encode information
- S_{vv} and S_{hh} do not present a zero response for general scatterer
- S_{vh} presents 2 zero response at $\theta = 0^{\circ}$ and $\theta = 90^{\circ}$ due to the symmetry of the loop
- $S_{vv}(\theta) = S_{hh}(\theta + \pi/2)$

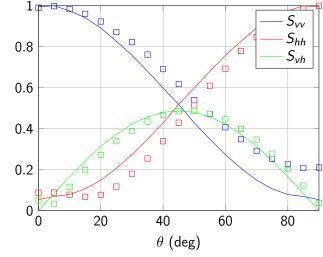
Measurement Bench



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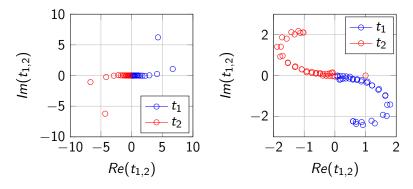
Normalized Scattering Parameter



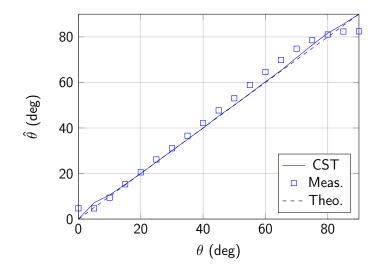
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Results



- Comparison of $t_{1,2}$ in simulation and measurement
- t_{1,2} should lie on the real axis
- Imaginary parts seems to be opposite...



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- Reference measurement has been chosen at $\theta = 0^{\circ}$
- Range of the sensor is $\hat{\theta} \in [0^\circ; 90^\circ]$
- $\bullet\,$ Average value of the absolute orientation error is $0.53^\circ\,$ in simulation
- $\bullet\,$ Average value of the absolute orientation error is $2.70^\circ\,$ in anechoic chamber

In this paper, we have presented a method which allows to extract the orientation of any chipless tag based on the knowledge of its polarization scattering matrix and a measurement of its co-polarization response in an unknown orientation.

- 3 reference measurements in known orientation
- a single measurement in co-polarization in unknown orientation
- $\bullet\,$ Error in anechoic chamber is less than 3° over a 90° interval
- Sensing is based in on amplitude and phase variation
- Identification can still be realized without reducing the coding capacity
- Method can be easily extended to cross-polariation case

Thank you for your kind attention

Feel free to send your questions/remarks to:

nicolas.barbot@lcis.grenoble-inp.fr